Modeling and Control of Complex Nonlinear Engineering Systems

Final Exam 2021-2022, Monday 11 April 2022, 8:30 - 10:30

- The exam is **open book**, meaning that the use of the course reader(s) as well as your written notes is allowed. The use of electronic devices is not allowed.
- The exam comprises four problems. Note that Problem 4 comes in **two versions**: please answer the problem according to your educational program (IEM/ME versus AM).
- Different from usual, the exam is for **two hours**.

Problem 1

(8+10+4+8=30 points)

(5 + 5 = 10 points)

Consider the nonlinear system

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} x_2 + x_2^2 + x_3^2 \\ x_3 + \sin(x_1 - x_3) \\ x_3^2 \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \\ 1 \end{bmatrix} u.$$
(1)

- (a) Characterize the points x_0 for which the system is locally strongly accessible at x_0 .
- (b) Is the system feedback linearizable to a controllable linear system?

For the remainder of this problem, consider the output

$$y = x_2. (2)$$

- (c) What is the relative degree of (1) with output (2)?
- (d) Using input-output linearization with output (2), determine a control law that achieves tracking of the reference trajectory

$$y_{\rm ref}(t) = \cos(\omega t)$$
 for some $\omega > 0$.

Problem 2

Consider the scalar nonlinear system

$$\dot{x} = x^2 + u. \tag{3}$$

To achieve feedback linearization, one can propose the feedback

$$u = -x^2 - x + v, \tag{4}$$

where v is a virtual input. For v = 0, the closed-loop system $\dot{x} = -x$ resulting from (3) and (4) has a globally asymptotically stable equilibrium at x = 0.

However, it turns out that, due to modelling errors, the actual nonlinear system reads

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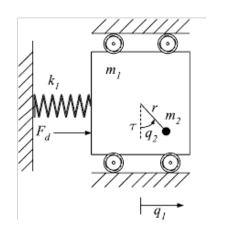
$$\dot{x} = (1+\varepsilon)x^2 + u,\tag{5}$$

for some small but nonzero ε .

- (a) Show that, for the perturbed closed-loop system resulting from (5) and (4), there exist initial conditions x_0 for which the resulting trajectory x(t) grows without bound (i.e., the perturbed closed-loop system does not have a globally asymptotically stable equilibrium).
- (b) Is it possible to guarantee stability of the origin for the perturbed closed-loop system by using a linearizing feedback different from (4)? If so, give such feedback. Note that we assume that ε is unknown.

Problem 3

Consider a translational oscillator with rotational actuator (TORA) system depicted below.



 q_1 : horizontal displacement of the cart

 q_2 : angular displacement of the pendulum

 m_1 : mass of the cart

 m_2 : mass of the pendulum

I: moment of inertia of m_2

- k_1 : spring constant
- τ : input torque
- F_d : external force

The total energy of the system is given by

$$H(q_1, q_2, p_1, p_2) = \frac{1}{2} \begin{bmatrix} p_1 \\ p_2 \end{bmatrix}^{\mathrm{T}} M(q_2)^{-1} \begin{bmatrix} p_1 \\ p_2 \end{bmatrix} + \frac{1}{2} k_1 q_1^2$$

with

$$M(q_2) = \begin{bmatrix} m_1 + m_2 & -m_2 r \cos q_2 \\ -m_2 r \cos q_2 & I + m_2 r^2 \end{bmatrix},$$

and where p_1 and p_2 are the translational and rotational momenta variables, respectively. The state space equations of this system can be written in the short-hand form

$$\dot{q} = \frac{\partial H}{\partial p}(q, p),$$

$$\dot{p} = -\frac{\partial H}{\partial q}(q, p) + u,$$
(6)

with

$$u = \begin{bmatrix} F_d \\ \tau \end{bmatrix}, \quad q = \begin{bmatrix} q_1 \\ q_2 \end{bmatrix}, \quad p = \begin{bmatrix} p_1 \\ p_2 \end{bmatrix}$$

Note that $\dot{q} = M(q_2)^{-1}p$. For reference, the full state space equations are given by

$$\begin{split} \dot{q}_1 &= \frac{(I+m_2r^2)p_1 + (m_2r\cos q_2)p_2}{(m_1+m_2)(I+m_2r^2) - m^2r^2\cos^2 q_2} \\ \dot{q}_2 &= \frac{(m_2r\cos q_2)p_1 + (m_1+m_2)p_2}{(m_1+m_2)(I+m_2r^2) - m^2r^2\cos^2 q_2} \\ \dot{p}_1 &= -k_1x_1 + F_d \\ \dot{p}_2 &= -\frac{\partial}{\partial q_2} \left(\frac{(I+m_2r^2)p_1^2 + 2(m_2r\cos q_2)p_1p_2 + (m_1+m_2)p_2^2}{(m_1+m_2)(I+m_2r^2) - m^2r^2\cos^2 q_2} \right) + \tau \end{split}$$

Hint: derivations in the following questions are easier using the short-hand form (6).

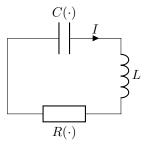
(a) Use the total energy H to determine the passive output corresponding to the input u.

(b) Introduce the feedback control

$$u = -\begin{bmatrix} d_1 & 0\\ 0 & d_2 \end{bmatrix} M(q_2)^{-1} p + v,$$

with $d_1 \ge 0$, $d_2 \ge 0$ and where $v = [v_1 \ v_2]^T$ is a new input. Is H still a storage function for showing that the input-output pair of (a) is passive for the closed loop system? Please motivate your answer.

(5+7+5+8+5=30 points)



Consider the RLC circuit above with nonlinear capacitor and resistor. It can be modelled as

$$\begin{bmatrix} \dot{Q} \\ \dot{I} \end{bmatrix} = \begin{bmatrix} I \\ -L^{-1}C(Q) - L^{-1}R(I) \end{bmatrix},\tag{7}$$

where Q is the charge of the capacitor and I the current through the network; L > 0 is the inductance constant. Moreover, the nonlinear capacitance function $C : \mathbb{R} \to \mathbb{R}$ and resistance function $R : \mathbb{R} \to \mathbb{R}$ are continuous and assumed to satisfy

$$\begin{aligned} QC(Q) &> 0 \qquad \forall \, Q \neq 0, \\ IR(I) &> 0 \qquad \forall \, I \neq 0. \end{aligned}$$

In the remainder of this problem, we will use the notation $x = \begin{bmatrix} Q & I \end{bmatrix}^{\mathrm{T}}$.

(a) Show that x = 0 is the unique equilibrium of (7).

As we are interested in stability of the equilibrium x = 0, introduce the function

$$V(x) = \frac{1}{2}LI^2 + \int_0^Q C(q) \,\mathrm{d}q.$$
 (8)

- (b) Show that V qualifies as a suitable Lyapunov function candidate.
- (c) Using V, prove that the equilibrium x = 0 is stable (in the sense of Lyapunov).
- (d) Is the equilibrium x = 0 also asymptotically stable?
- (e) Provide a condition (in terms of the functions C and/or R) under which the equilibrium x = 0 is globally asymptotically stable.

Problem 4 (for AM students)

Consider N nonlinear systems of the form

$$\dot{x}_i = f_i(x_i, u_i),$$

$$y_i = h_i(x_i),$$

with state $x_i(t) \in \mathbb{R}^{n_i}$ and scalar input $u_i(t) \in \mathbb{R}$ and output $y_i(t) \in \mathbb{R}$, with i = 1, 2, ..., N. In addition, $f_i(0,0) = 0$ and $h_i(0) = 0$ hold.

For each system *i*, we assume furthermore that it is zero-state observable and has an L_2 -gain bounded by γ_i , i.e., it is dissipative with respect to the supply rate

$$s_i(u_i, y_i) = \frac{1}{2}\gamma_i^2 ||u_i||^2 - \frac{1}{2} ||y_i||^2$$

for some differentiable storage function S_i . We collect states, inputs, outputs, and L_2 gain bounds in the vectors and diagonal matrix

$$x = \begin{bmatrix} x_1 \\ \vdots \\ x_N \end{bmatrix}, \qquad u = \begin{bmatrix} u_1 \\ \vdots \\ u_N \end{bmatrix}, \qquad y = \begin{bmatrix} y_1 \\ \vdots \\ y_N \end{bmatrix}, \qquad \Gamma = \begin{bmatrix} \gamma_1 & 0 \\ \ddots \\ 0 & \gamma_N \end{bmatrix},$$

respectively, and assume that the N nonlinear systems are interconnected as

$$u = Ky$$

for some matrix $K \in \mathbb{R}^{N \times N}$.

(a) Assume that there exists a *diagonal* positive definite matrix P such that

$$(\Gamma K)^{\mathrm{T}} P(\Gamma K) - P \prec 0, \tag{9}$$

where $X \prec 0$ indicates that the matrix X is negative definite. Show that the origin x = 0 is an asymptotically stable equilibrium point for the interconnected system. *Hint.* Write

$$P = \begin{bmatrix} p_1 & 0 \\ & \ddots \\ 0 & p_N \end{bmatrix}$$

and consider the function $V(x) = \sum_{i=1}^{N} p_i S_i(x_i)$.

In the remainder of this problem, we extend the interconnection to include an external input v and external output z as

$$u = Ky + Lv, \quad z = My.$$

(b) Using the same function V as in (a) as a storage function, give a sufficient condition for guaranteeing that the interconnected system is dissipativive with respect to

$$s(v,z) = \frac{1}{2}\theta^2 ||v||^2 - \frac{1}{2}||z||^2$$

i.e., has an L_2 -gain bounded by θ with respect to the external input v and output z. Specifically, give the sufficient condition in terms of a matrix inequality in a similar spirit as (9).

 $^{(10 \}text{ points free})$